Gravitational wave data analysis: Detection

Gravitational waves

Linearized general relativity: Einstein equations become a wave equation

$$\Box \bar{h}_{\mu\nu} = -\frac{16\pi G}{c^4} T_{\mu\nu}$$

• Gravitational radiation is primarily generated by time-dependent mass quadrupole moment:

$$\left[h_{ij}^{\mathrm{TT}}(t,\mathbf{x})\right]_{\mathrm{quad}} = \frac{1}{r} \frac{2G}{c^2} \Lambda_{ij,kl}(\hat{\mathbf{n}}) \, \ddot{M}^{kl}(t - r/c)$$

- Nearby point particles in free fall, small separation ζ^{μ}
 - → Tidal effect:

$$\ddot{\zeta}^i = \frac{1}{2} \ddot{h}_{ij}^{\rm TT} \zeta^j$$

 $h_{_{\scriptscriptstyle +}}$

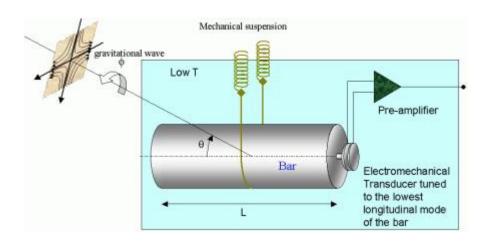
h

"plus" and "cross" polarizations:

$$\ddot{\zeta}^i = \frac{1}{2} \ddot{h}_{ij}^{\rm TT} \zeta^j$$

Exploit tidal effect on matter

- Resonant detectors
 - Metal bar: resonant frequency
 - Measure tiny length changes
 - Also spherical resonant detectors:
 Equal sensitivity in all directions (miniGRAIL, Leiden)

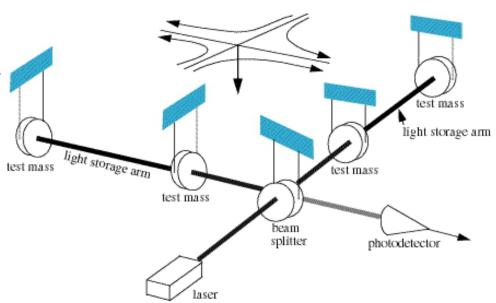




Interferometric detectors

$$\ddot{\zeta}^i = \frac{1}{2} \ddot{h}_{ij}^{\rm TT} \zeta^j$$

- Interferometric detectors
 - Laser beam through beam splitter
 - Power built up in long (several km) cavities
 - At output: destructive interference
 - ... unless gravitational wave present



Interferometric detectors

$$\ddot{\zeta}^i = \frac{1}{2} \ddot{h}_{ij}^{\rm TT} \zeta^j$$

Motion of the end mirrors

$$\delta \ddot{\zeta}_1^x = \frac{1}{2} \ddot{h}_{xx} (L + \delta \zeta_1^x)$$

$$\delta \ddot{\zeta}_2^y = \frac{1}{2} \ddot{h}_{yy} (L + \delta \zeta_2^y)$$

... hence

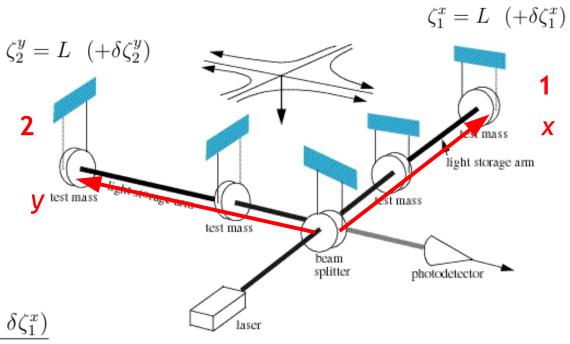
$$\delta \zeta_1^x = \frac{1}{2} h_{xx} L$$

$$\delta \zeta_2^y = \frac{1}{2} h_{yy} L$$

Measured strain:

$$h(t) \equiv \frac{\Delta L}{L} = \frac{(L + \delta \zeta_2^y) - (L + \delta \zeta_1^x)}{L}$$

$$h(t) = \frac{1}{2}(h_{xx} - h_{yy})$$



For L-shaped detectors:

$$h(t) = \frac{1}{2}(h_{xx} - h_{yy})$$

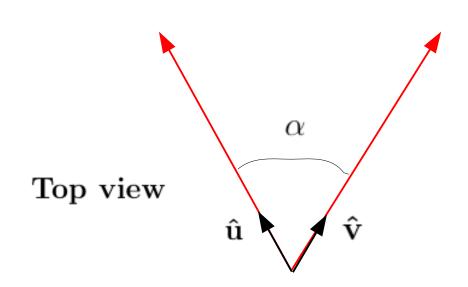
More generally:

$$h(t) = \frac{1}{2}(h_{uu} - h_{vv})$$

Detector tensor:

$$D^{ij} = \frac{1}{2}(u^i u^j - v^i v^j)$$

$$h(t) = D^{ij}h_{ij}$$



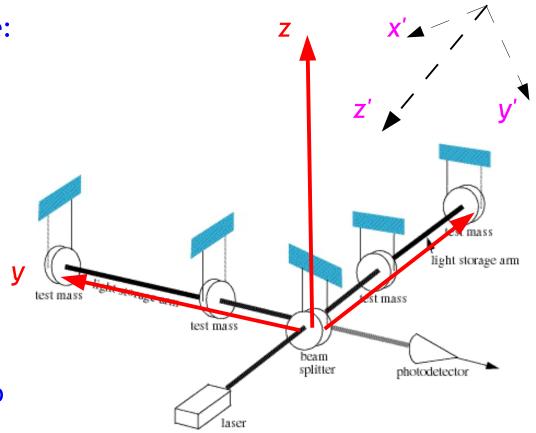
$$h(t) = \frac{1}{2}(h_{xx} - h_{yy})$$

Signal from arbitrary direction

In transverse-traceless frame:

$$h_{ij}^{\hat{\mathbf{n}} = \hat{\mathbf{z}}'} = \begin{pmatrix} h_{+} & h_{\times} & 0\\ h_{\times} & -h_{+} & 0\\ 0 & 0 & 0 \end{pmatrix}_{ij}$$

- If x' = x, y' = y, z' = z then $h(t) = h_+(t)$
- In general, need to apply linear transformation $h_{ij}^{\mathbf{det}} = \mathcal{R}_{ik} \mathcal{R}_{jl} h_{kl}^{\hat{\mathbf{n}} = \hat{\mathbf{z}}'}$
- Projection onto detector also linear, hence



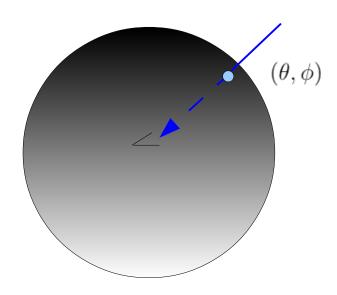
$$h(t) = D^{ij} h_{ij}^{\mathbf{det}} = D^{ij} \mathcal{R}_{ik} \mathcal{R}_{jl} h_{kl}^{\hat{\mathbf{n}} = \hat{\mathbf{z}}'} = F_+ h_+(t) + F_\times h_\times(t)$$

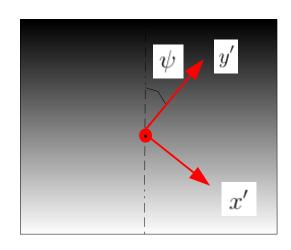
$$h(t) = F_+ h_+(t) + F_\times h_\times(t)$$

For an L-shaped interferometer:

$$F_{+} = \frac{1}{2}(1 + \cos^{2}(\theta)) \cos(2\phi) \cos(2\psi) - \cos(\theta) \sin(2\phi) \sin(2\psi)$$

$$F_{\times} = \frac{1}{2}(1 + \cos^{2}(\theta)) \cos(2\phi) \sin(2\psi) + \cos(\theta) \sin(2\phi) \cos(2\psi)$$





ullet $(heta,\phi)$ sky position; ψ polarization angle

Digging a signal out of noise

If there is a signal: measured strain is sum of noise and signal:

$$s(t) = n(t) + h(t)$$

If shape of signal approximately known: integrate against the output

$$\frac{1}{T} \int_0^T s(t) \, h(t) \, dt = \frac{1}{T} \int_0^T n(t) \, h(t) \, dt + \frac{1}{T} \int_0^T h(t)^2 dt$$
 oscillatory positive definite
$$\sim \left(\frac{\tau_0}{T}\right)^{1/2} n_0 h_0 \qquad \sim h_0^2$$

 τ_0 period of the GW signal,

 h_0 characteristic signal amplitude

 n_0 characteristic amplitude of the noise

- lacktriangledown To detect the signal, don't need $h_0>n_0$ but only $h_0>(au_0/T)^{1/2}n_0$
- lacktriangle Binary objects: $au_0 \sim 10^{-2}\,\mathrm{s}$, $T \sim 100\,\mathrm{s}$ o $(au_0/T)^{1/2} \sim 10^{-2}$
- Millisecond pulsars: $au_0=1\,\mathrm{ms}$, $T\sim 1\,\mathrm{yr}$ o $(au_0/T)^{1/2}\sim 10^{-5}$

Characterizing the noise

- Detector data comes in as time series
- If only noise:

$$(n(t_0), n(t_1), \dots, n(t_N))$$
 where $t_{i+1} = t_i + \Delta t$

Often convenient to take a (discrete) Fourier transform:

$$(\tilde{n}(f_0), \tilde{n}(f_1), \dots, \tilde{n}(f_N))$$
 where $f_{i+1} = f_i + \Delta f$
Notation: $\tilde{n}(f_i) = \tilde{n}_i$

- Some noise realizations are more probable than others Probability distribution in each frequency bin: $p(\tilde{n}_i)$
- We will assume that the noise is *stationary* and *Gaussian*:

$$p(ilde{n}_i) \propto \, e^{-rac{| ilde{n}_i|^2}{2\sigma_i^2}}$$

Stationarity and Gaussianity:

$$\langle \tilde{n}_i \rangle = \int \tilde{n}_i \, p(\tilde{n}_i) \, d\tilde{n}_i = 0 \qquad \qquad \langle |\tilde{n}_i|^2 \rangle = \int |\tilde{n}_i|^2 \, p(\tilde{n}_i) \, d\tilde{n}_i = \sigma_i^2$$

Probability density for noise realization as a whole:

$$p[n] = p(\tilde{n}_0, \tilde{n}_1, \dots, \tilde{n}_N) = \prod_{i=1}^N p(\tilde{n}_i)$$

Characterizing the noise

Probability density for noise realization as a whole:

$$p[n] = p(ilde{n}_0, ilde{n}_1, \ldots, ilde{n}_N) = \prod_{i=1}^N p(ilde{n}_i)$$

For stationary, Gaussian noise:

$$p[n] = \prod_{i=1}^N p(ilde{n}_i) = \mathcal{N}\,e^{-rac{1}{2}\sum_{i=1}^N rac{| ilde{n}_i|^2}{\sigma_i^2}}$$

For purpose of this lecture, convenient to take continuum limit:

$$p[n] = \mathcal{N}e^{-\frac{1}{2}\sum_{i=1}^{N}\frac{|\tilde{n}_{i}|^{2}}{\sigma_{i}^{2}}}$$

$$= \mathcal{N}e^{-\frac{1}{2}\sum_{i=1}^{N}\frac{|\tilde{n}_{i}|^{2}}{\sigma_{i}^{2}\Delta f}\Delta f}$$

$$\to \mathcal{N}e^{-\frac{1}{2}\int_{-\infty}^{\infty}\frac{|\tilde{n}(f)|^{2}}{\frac{1}{2}S_{n}(f)}df}$$

Variance:

$$\sigma_i^2 \Delta f \to \frac{1}{2} S_n(f)$$
 or $\sigma_i^2 \to \underbrace{\delta(f - f')}_{\sim \frac{1}{\Delta f}} \underbrace{\frac{1}{2} S_n(f)}$

$$\langle \tilde{n}^*(f) \, \tilde{n}(f') \rangle = \delta(f - f') \, \frac{1}{2} \, S_n(f)$$

Characterizing the noise

Probability density and variance for noise realizations:

$$p[n] = \mathcal{N} e^{-\frac{1}{2} \int_{-\infty}^{\infty} \frac{|\tilde{n}(f)|^2}{\frac{1}{2} S_n(f)} df}$$
$$\langle \tilde{n}^*(f) \, \tilde{n}(f') \rangle = \delta(f - f') \, \frac{1}{2} \, S_n(f)$$

- Could also have worked in time domain
 - Stationarity: $\langle n(t) \rangle = 0$
 - Gaussianity: completely determined by $R(\tau) \equiv \langle n(t+\tau) \, n(t) \rangle$ where again $\langle \dots \rangle$ denotes average over noise realizations Defining noise power spectral density as

$$\frac{1}{2}S_n(f) \equiv \int_{-\infty}^{\infty} d\tau \, R(\tau) \, e^{2\pi i f \tau}$$

one finds

$$\langle \tilde{n}^*(f)\,\tilde{n}(f')\rangle = \frac{1}{2}\delta(f-f')\,S_n(f)$$

Instead of integrating the data against waveforms, use more generic filter:

$$\hat{s} = \int_{-\infty}^{\infty} dt \, s(t) \, \underbrace{K(t)}_{filter}$$

• Define S to be the expected value if a signal h(t) is present, and N the root-mean-square value if no signal present:

$$S = \langle \hat{s} \rangle_h \qquad N = \left[\langle \hat{s}^2 \rangle_{h=0} - \langle \hat{s} \rangle_{h=0}^2 \right]^{1/2}$$

Define signal-to-noise ratio:

$$\frac{S}{N} = \frac{\langle \hat{s} \rangle_h}{\left[\langle \hat{s}^2 \rangle_{h=0} - \langle \hat{s} \rangle_{h=0}^2 \right]^{1/2}}$$

Now find out which filter K(t) maximizes S/N

$$\frac{S}{N} = \frac{\langle \hat{s} \rangle_h}{\left[\langle \hat{s}^2 \rangle_{h=0} - \langle \hat{s} \rangle_{h=0}^2 \right]^{1/2}}$$

$$\hat{s} = \int_{-\infty}^{\infty} dt \, s(t) \, \underbrace{K(t)}_{filter}$$

Write 5 in the frequency domain:

$$S = \langle \hat{s} \rangle_h$$

$$= \int_{-\infty}^{\infty} dt \, \langle s(t) \rangle \, K(t)$$

$$= \int_{-\infty}^{\infty} dt \, h(t) \, K(t)$$

$$= \int_{-\infty}^{\infty} df \, \tilde{h}(f) \, \tilde{K}^*(f)$$

... and *N*:

$$N = \left[\langle \hat{s}^2 \rangle - \langle \hat{s} \rangle^2 \right]_{h=0}^{1/2}$$

$$= \left[\langle \hat{s}^2 \rangle \right]_{h=0}^{1/2}$$

$$= \left[\int_{-\infty}^{\infty} dt \int_{-\infty}^{\infty} dt' \left\langle n(t) \, n(t') \right\rangle K(t) \, K(t') \right]^{1/2}$$

$$= \left[\int_{-\infty}^{\infty} df \, \frac{1}{2} S_n(f) \, |\tilde{K}(f)| \right]^{1/2}$$

... to arrive at:

$$\frac{S}{N} = \frac{\int_{-\infty}^{\infty} df \, \tilde{h}(f) \, \tilde{K}^*(f)}{\left[\int_{-\infty}^{\infty} df \, \frac{1}{2} S_n(f) \, |\tilde{K}(f)|^2 \right]^{1/2}}$$

Now define the noise-weighted inner product

$$(A|B) \equiv \operatorname{Re} \int_{-\infty}^{\infty} df \frac{\tilde{A}^{*}(f) \, \tilde{B}(f)}{\frac{1}{2} S_{n}(f)}$$
$$= 4 \operatorname{Re} \int_{0}^{\infty} \frac{\tilde{A}^{*}(f) \, \tilde{B}(f)}{S_{n}(f)}$$

... and rewrite S/N as

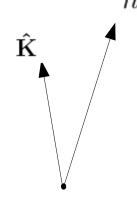
$$\frac{S}{N} = \frac{(\mathbf{K}|h)}{(\mathbf{K}|\mathbf{K})^{1/2}}$$

$$\mathbf{K} = \frac{1}{2} S_n(f) \, \tilde{K}(f)$$

... or

$$\frac{S}{N} = (\hat{\mathbf{K}}|h)$$

$$\hat{\mathbf{K}} = rac{\mathbf{K}}{(\mathbf{K}|\mathbf{K})^{1/2}}$$



Maximizing S/N is equivalent to making $\hat{\mathbf{K}}$ point in the same direction as h !

<u>Matched filtering</u>

$$\frac{S}{N} = (\hat{\mathbf{K}}|h)$$

$$\hat{\mathbf{K}} = \frac{\mathbf{K}}{(\mathbf{K}|\mathbf{K})^{1/2}}$$

$$\frac{S}{N} = (\hat{\mathbf{K}}|h) \qquad \qquad \hat{\mathbf{K}} = \frac{\mathbf{K}}{(\mathbf{K}|\mathbf{K})^{1/2}} \qquad \qquad \mathbf{K} = \frac{1}{2}S_n(f)\,\tilde{K}(f)$$

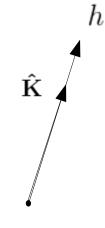
lacktriangle Need to make $\hat{\mathbf{K}}$ point in the same direction as h

$$\hat{\mathbf{K}} \propto h$$

 $\mathbf{K} \propto h$



$$\tilde{K}(f) \propto \frac{\tilde{h}(f)}{S_n(f)}$$



"Wiener filter"

Substituting back into the expression for S/N gives the highest signal-to-noise ratio one can attain:

$$\frac{S}{N} = \frac{(h|\langle s \rangle)}{\sqrt{(h|h)}}$$

$$= \left[4 \int_0^\infty df \, \frac{|\tilde{h}(f)|^2}{S_n(f)}\right]^{1/2}$$

$$\frac{S}{N} = \frac{(h|\langle s \rangle)}{\sqrt{(h|h)}}$$

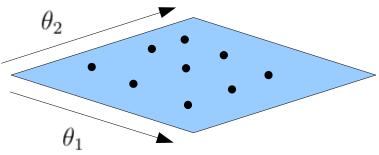
$$= \left[4 \int_0^\infty df \, \frac{|\tilde{h}(f)|^2}{S_n(f)}\right]^{1/2}$$

- In practice:
 - No access to *expected* s(t), only the actual detector output s(t)
 - Waveforms are characterized by source parameters $\bar{\theta} = \{\theta_1, \theta_2, \dots, \theta_N\}$
 - Use optimal filter for many choices of parameters and find largest S/N:

$$\frac{S}{N} = \max_{\bar{\theta}} \frac{(h(\bar{\theta})|s)}{\sqrt{(h(\bar{\theta})|h(\bar{\theta}))}}$$

- Parameter space can only be sampled in a discrete manner:

"template banks"



- Noise not perfectly Gaussian or stationary

Matched filtering in practice: compact binary coalescence

- Simplest compact binary coalescence waveforms:
 - Inspiral only
 - Amplitude to leading PN order only

$$h_{+} = h_0 \cos (2\varphi(t))$$
$$h_{\times} = h_{\frac{\pi}{2}} \sin (2\varphi(t))$$

Detector response:

$$h(t) = F_+ h_+ + F_\times h_\times$$

Can be written as:

$$h(t) = A(t)\cos(2\varphi(t))\cos(\Phi_0) + A(t)\sin(2\varphi(t))\sin(\Phi_0)$$

where

$$A(t) = \left[F_+^2 h_0^2 + F_{\times}^2 h_{\frac{\pi}{2}}^2 \right]^{1/2}$$
 $\cos \Phi_0 = \frac{F_+ h_0}{A},$
 $\sin \Phi_0 = \frac{F_{\times} h_{\frac{\pi}{2}}}{A}.$

Define

$$h_c = A(t)\cos 2\varphi(t)$$

 $h_s = A(t)\sin 2\varphi(t)$

Matched filtering in practice: compact binary coalescence

$$h(t) = A(t)\cos(2\varphi(t))\cos(\Phi_0) + A(t)\sin(2\varphi(t))\sin(\Phi_0)$$

$$h_c = A(t)\cos 2\varphi(t)$$

$$h_s = A(t)\sin 2\varphi(t)$$

• Normalize h_c , h_s :

$$\hat{h}_c = rac{h_c}{(h_c|h_c)} \quad \hat{h}_s = rac{h_s}{(h_s|h_s)}$$

Matched filtering against detector output s:

$$(s|\hat{h}) = (s|\hat{h}_c) \cos(\Phi_0) + (s|\hat{h}_s) \sin(\Phi_0)$$

= $\left[(s|\hat{h}_c)^2 + (s|\hat{h}_s) \right]^{1/2} \cos(\Phi_0 - \alpha)$

where $\alpha = \tan^{-1}\frac{(s|h_s)}{(s|\hat{h}_c)}$ and the S/N ratio is maximized when $\Phi_0 = \alpha$

Matched filtering reduces to:

$$(s|\hat{h}) = \left[(s|\hat{h}_c)^2 + (s|\hat{h}_s) \right]^{1/2}$$

No dependence of templates on extrinsic parameters! Only dependence on masses and spins

Placement of template banks

Match between nearby templates:

$$M = (\hat{h}(\bar{\theta})|\hat{h}(\bar{\theta} + \Delta\bar{\theta}))$$

Expand in small quantities:

$$M \simeq (\hat{h}(\bar{\theta})|\hat{h}(\bar{\theta})) + \frac{\partial M}{\partial \theta^{i}} \Delta \theta^{i} + \frac{1}{2} \frac{\partial^{2} M}{\partial \theta^{i} \partial \theta^{j}} \Delta \theta^{i} \Delta \theta^{j}$$
$$= 1 + \frac{1}{2} \frac{\partial^{2} M}{\partial \theta^{i} \partial \theta^{j}} \Delta \theta^{i} \Delta \theta^{j}$$

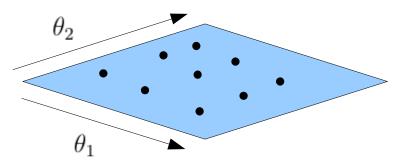
Define a metric tensor on parameter space:

$$g_{ij} \equiv -\frac{1}{2} \frac{\partial^2 M}{\partial \theta^i \partial \theta^j}$$

Mismatch between nearby templates:

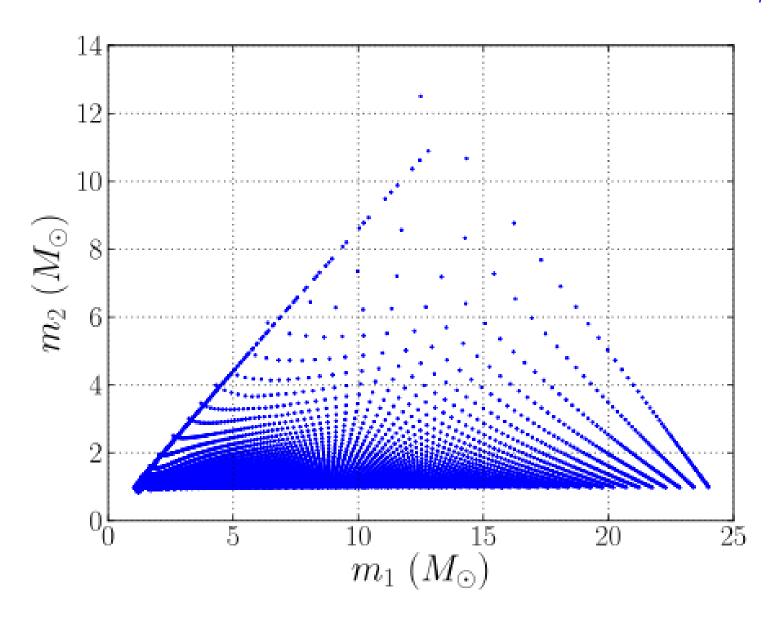
$$1 - M = g_{ij} \Delta \theta^i \Delta \theta^j$$

Place templates on parameter space such that metric distance never larger than a pre-specified mismatch, e.g. 0.03



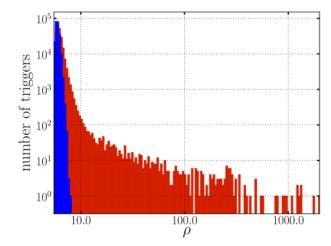
Placement of template banks

• Example where spins ignored so that templates only placed in (m_1, m_2) :

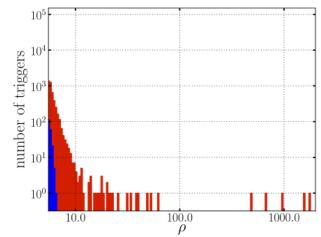


Handling of potential signals

- Glitches in the instruments can pose as gravitational wave signals!
- Distribution of SNRs for noise triggers in a single detector, in Gaussian noise
 (blue) and in real data (red)



 Demand that event seen in at least two detectors, with same parameters up to expected uncertainties (to be explained later)

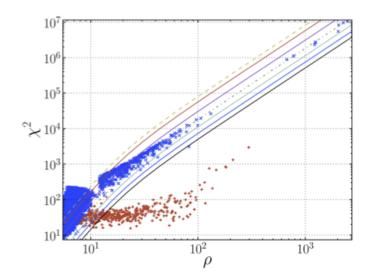


Signal consistency tests

- Demand that build-up of SNR over frequency is as expected if a signal is really present
 - Divide frequency range in n bins that should contribute equally to SNR $(h_k|h_k)=rac{(h|h)}{n}$
 - Construct χ^2 statistic

$$\chi^2 = \sum_{k=1}^n \left[\left((s|h_{c,k}) - \frac{(h_c|h_c)}{n} \right)^2 + \left((s|h_{s,k}) - \frac{(h_c|h_c)}{n} \right)^2 \right]$$

Noise transients will typically have high χ^2



Compare distribution of noise triggers (blue) and simulated GW signals (red)

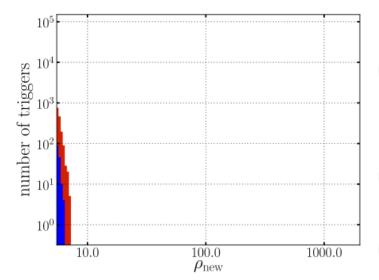
Signal consistency tests

• Construct a detection statistic that is large when SNR large and χ^2 small, e.g. "new SNR"

$$ho_{
m new} = \left\{ egin{array}{ll}
ho & {
m for} \ \chi^2 \leq n_{
m dof} \
ho \left[rac{1}{2} \left(1 + \left(rac{\chi^2}{n_{
m dof}}
ight)^3
ight)
ight]^{-1/6} & {
m for} \ \chi^2 > n_{
m dof} \ \end{array}
ight.$$

Distribution of new SNR for noise triggers in Gaussian noise (blue) and real

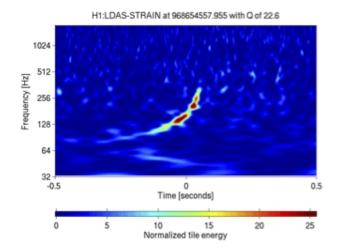
data (red)



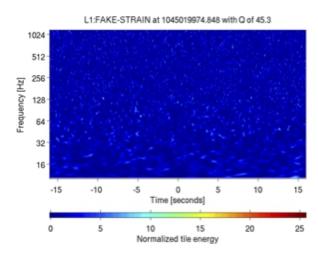
- Given data from multiple detectors, slide data stream in time w.r.t. each other
- Construct distribution of new SNR for multi-detector triggers (necessarily fake events!)
- Use this to set a threshold for a potential signal to overcome
- Given a candidate signal, assess its significance by comparing with this distribution

Other diagnostic tools

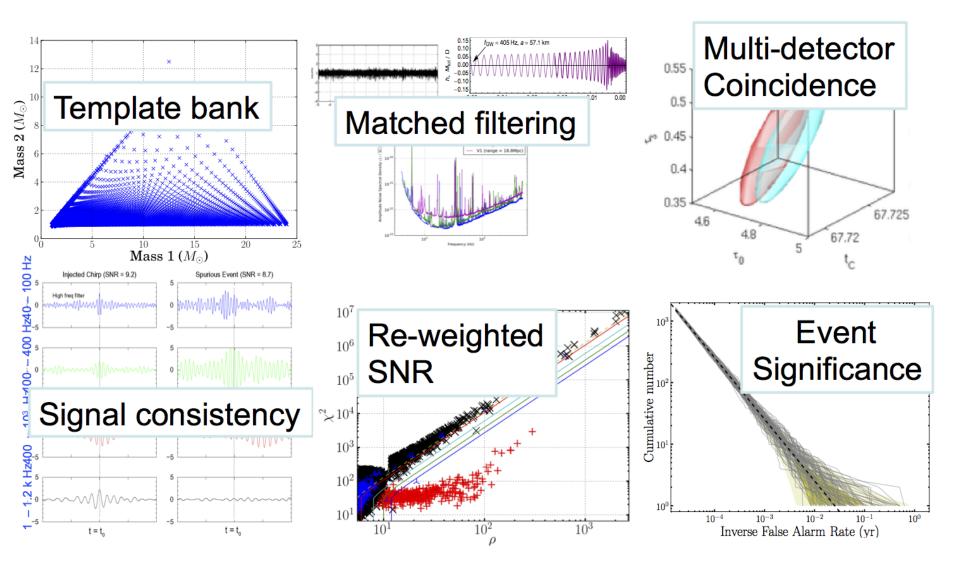
- Detector characterization and data quality studies
- Visual inspection of time-frequency spectrograms
 - Neutron star-black hole coalescence:



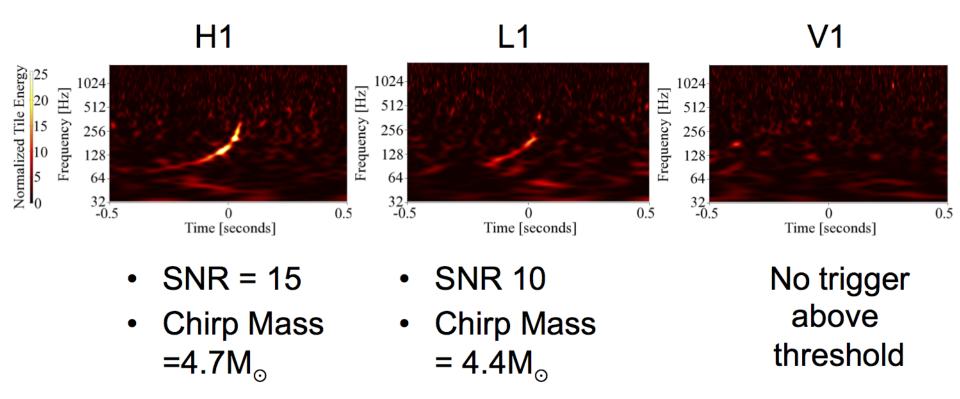
- Binary neutron star coalescence:



Summary so far

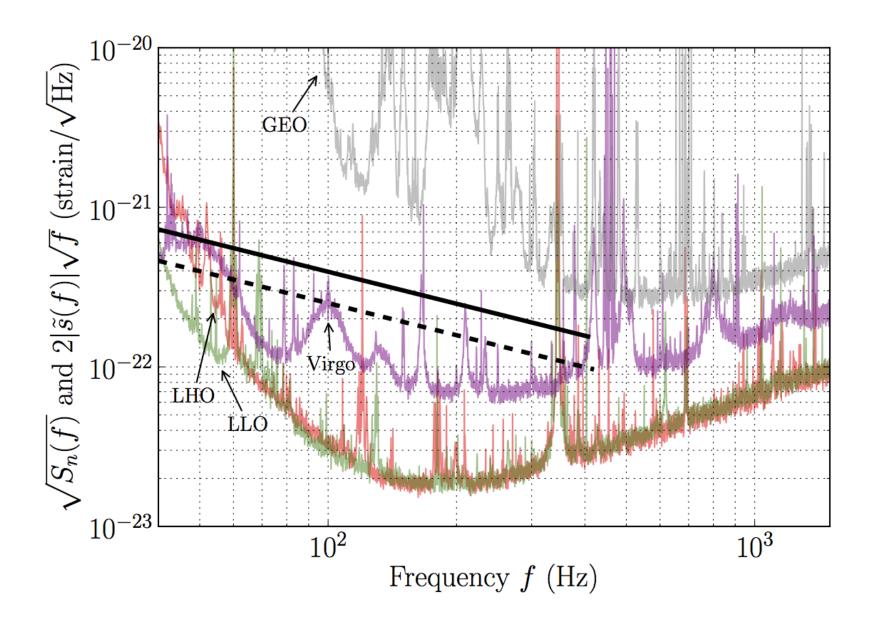


The Big Dog

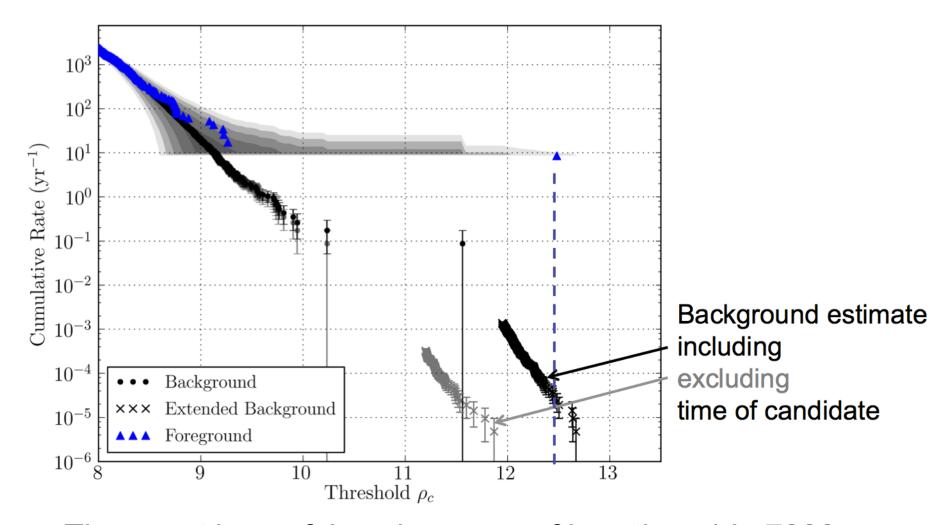


A coincident signal was observed by the two LIGO detectors at 2010 September 16, 06:42:23 UTC, with ρ_c = 12.5

The Big Dog



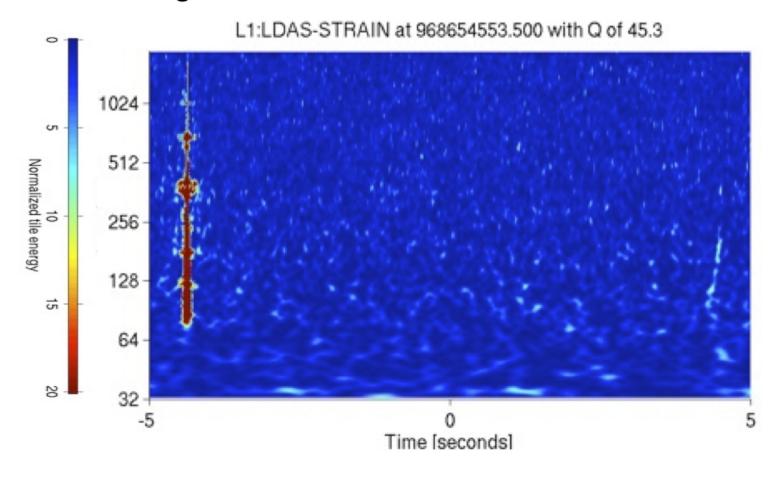
The Big Dog: significance of the event



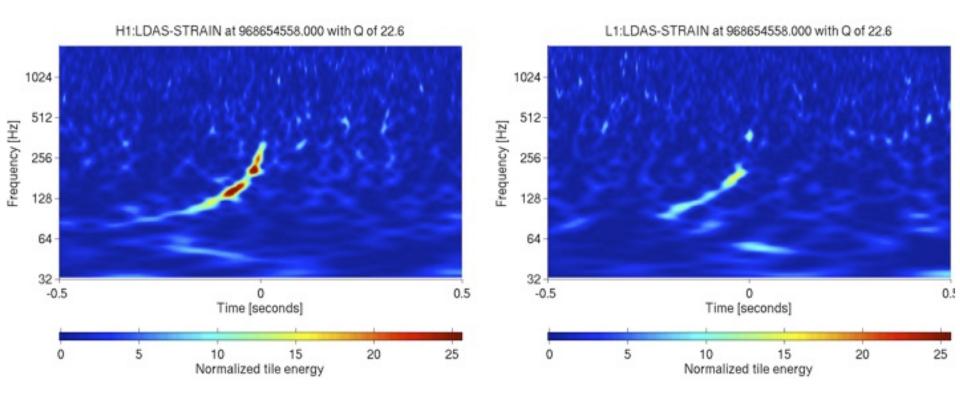
The event has a false alarm rate of less than 1 in 7000 y.

The Big Dog: instrumental validation

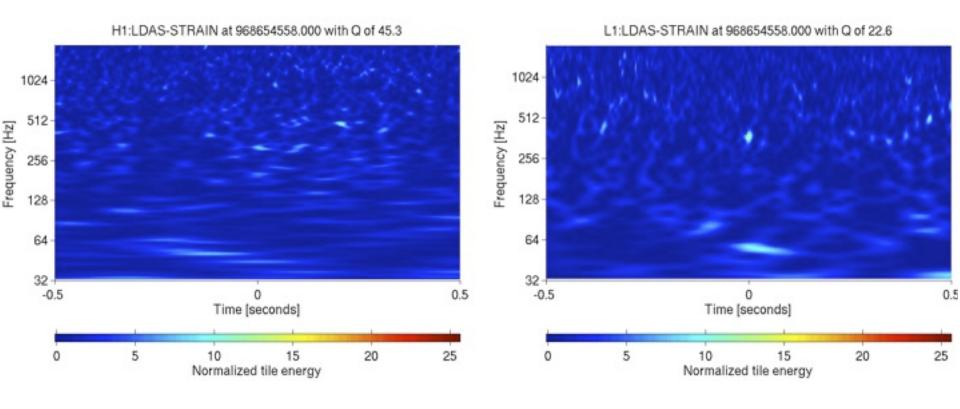
We find no evidence that the signal was of instrumental or environmental origin.



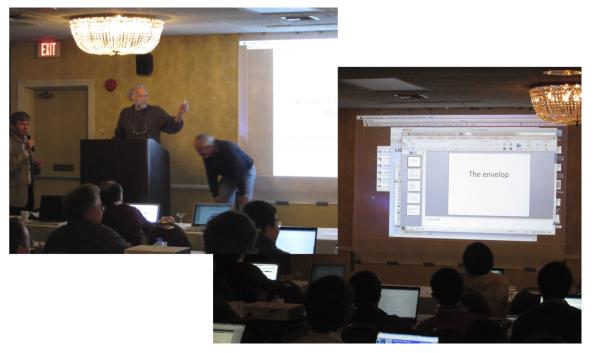
The Big Dog: waveform subtraction



The Big Dog: waveform subtraction



But this one wasn't real...



Blind hardware injection:

- Mirrors are shaken so as to mimick what a real signal would do
- Nobody is told!
- Check if the data analysis algorithms are working properly
- Plausible deniability...

Better luck in the advanced detector era!